

GRF5526

HIGH LINEARITY POWER AMPLIFIER 2.2 to 2.7 GHz

FEATURES

- Excellent OP1dB, OIP3, ACLR and IM3 Performance
- Native Linearity Provides up to +23 dBm P_{OUT} with > 45 dBc ACLR – Without the Need for Digital Predistortion Correction
- +23 dBm Linear Output Power Maintained at 105 °C
- Flexible Biasing Provides Latitude for Linearity Optimization
- 150 mA Native Mode Quiescent Current Consumption
- 5 V Supply Voltage
- 50 Ω Single-ended Input and Output Impedances
- Digital Shutdown
- Rugged Design is Extremely Resilient to Mismatched Loads
- -40 to 105 °C Operating Temperature Range
- Compact 3 x 3 mm QFN-16 Package

Reference: 5 V / 150 mA I_{CCQ} / 2.6 GHz

- Gain: 28.5 dB
- OIP3: 45 dBm @ 23 dBm $P_{OUT}/tone$
- OP1dB: 32 dBm
- Noise Figure: 3 dB

APPLICATIONS

- Cellular Boosters/Repeaters
- Automotive Compensators
- Picocells/Femtocells
- Customer Premise Equipment
- Cellular DAS
- Wireless Infrastructure

DESCRIPTION

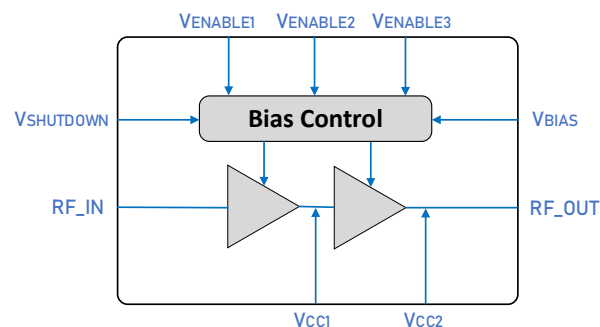
The GRF5526 is a high gain, two-stage InGaP HBT power amplifier designed to deliver excellent P1dB, ACLR and IM3 performance over the 2.2 to 2.7 GHz band. Its exceptional native linearity makes it an ideal choice for transmitter applications that typically do not employ digital predistortion correction schemes.

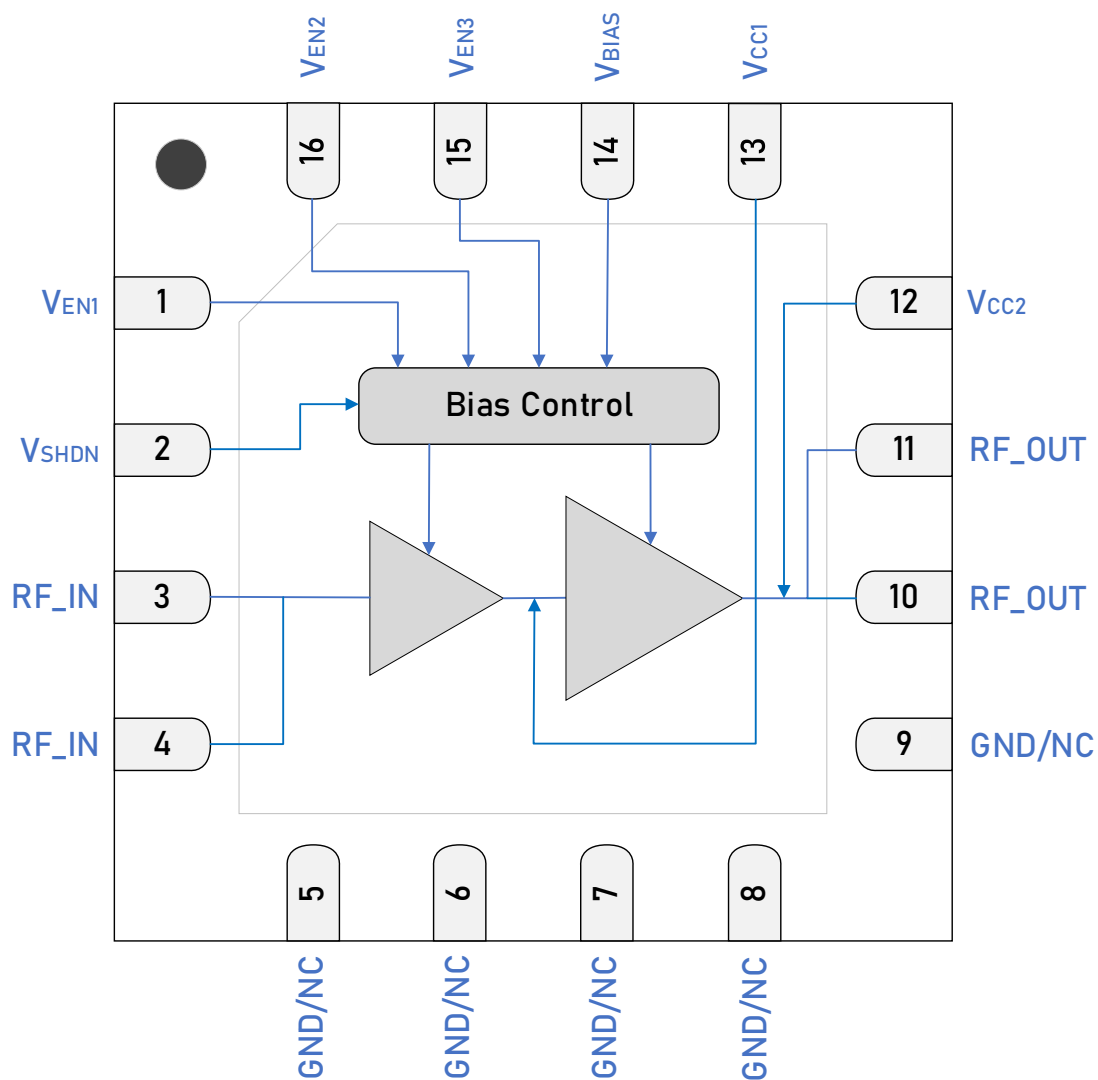
This device is part of a complete family of externally matched linear amplifiers that cover the following frequency ranges:

GRF5506: 0.66 - 0.72 GHz	GRF5518: 1.8 - 2.0 GHz
GRF5507: 0.7 - 0.91 GHz	GRF5519: 1.92 - 2.2 GHz
GRF5508: 0.777 - 0.96 GHz	GRF5521: 2.11 - 2.17 GHz
GRF5510: 0.88 - 0.96 GHz	GRF5526: 2.2 - 2.7 GHz
GRF5517: 1.6 - 1.92 GHz	GRF5536: 3.3 - 4.2 GHz

Please consult with the GRF applications engineering team for custom tuning/evaluation board data.

BLOCK DIAGRAM





3 x 3 mm QFN-16 Pin Out (Top View)

Pin Assignments

Pin	Name	Description	Note
1	V _{EN1}	Enable1 Voltage Input	V _{EN1} and series resistor set I _{CCQ} for the input stage. V _{EN1} ≤ 0.2 V disables stage 1.
2	V _{SHDN}	Digital Shutdown Pin	V _{SHDN} ≥ 1.7 V (Logic HIGH) disables device. V _{SHDN} ≤ 0.9 V (Logic LOW) enables device.
3, 4	RF_IN	RF Input	Pins 3 & 4 tied together on system board. An external DC blocking cap must be used.
5, 6, 7, 8, 9	GND/NC	Ground or No Connect	No internal connection to die. These pins can be left unconnected, or be connected to ground (recommended). Use a via as close to the pin as possible if grounded.
10, 11	RF_OUT	PA Output	Pins 10 & 11 tied together on system board.
12	V _{CC2}	Bias Voltage	V _{CC2} must be applied to this pin via an RF choke.
13	V _{CC1}	Bias Voltage	V _{CC1} must be applied to this pin via L-C interstage match.
14	V _{BIAS}	Bias Circuit Supply	Connect to V _{CC2} through external resistor.
15	V _{EN3}	Enable3 Voltage Input	Bias Voltage applied via series resistor.
16	V _{EN2}	Enable2 Voltage Input	V _{EN2} and series resistor set I _{CCQ} for the output stage. V _{EN2} ≤ 0.2 V disables stage 2.
PKG BASE	GND	Ground	Provides DC and RF ground for the amplifier, as well as thermal heat sink. Recommend multiple 8 mil vias beneath the package for optimal RF and thermal performance. Refer to evaluation board top layer graphic on schematic page.

Note 1: GRF5526 and GRF5536 have a common pinout that is slightly different from the pinout shared by all the other GRF55XX PAs. Please see page 2.

Truth Table

Pin	Logic	Condition
V _{SHDN}	LOW	Full Operation
	HIGH	All Amplifiers Off
V _{EN1}	LOW	Stage 1 Amplifier Off
	HIGH	Stage 1 Amplifier On
V _{EN2}	LOW	Stage 2 Amplifier Off
	HIGH	Stage 2 Amplifier On
V _{EN3}	LOW	Stage 2 Amplifier Off
	HIGH	Stage 2 Amplifier On

Absolute Ratings

Parameter		Symbol	Min.	Max.	Unit
Supply Voltage		V_{CC}	4	5.5	V
RF Input Power	50 Ω , $V_{CC} = 5$ V, CW Tone, 100% Duty Cycle, $T_{PKG\ BASE} = 25$ °C	$P_{IN\ MAX - 1:1}$		20	dBm
	Load VSWR $\leq 8:1$, all phase angles, $V_{CC} = 5$ V, CW Tone, 100% Duty Cycle, $T_{PKG\ BASE} = -40$ to 105 °C	$P_{IN\ MAX - 8:1}$		17	
Operating Temperature (package base)		$T_{PKG\ BASE}$	-40	105	°C
Maximum Junction Temperature (MTTF > 10 ⁶ hours)		$T_{J\ MAX}$		170	°C
Maximum Dissipated Power (Stage 1). DC only (no RF applied).		$P_{DISS\ MAX}$		* 300	mW
Maximum Dissipated Power (Stage 2). DC only (no RF applied).		$P_{DISS\ MAX}$		* 1000	mW
Shutdown Voltage		V_{SHDN}		** 5.5	V

Electrostatic Discharge

Human Body Model	HBM	500		V
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Storage

Storage Temperature	T_{STG}	-65	150	°C
Moisture Sensitivity Level	MSL		1	--

* Bias resistors M2/M4 have been empirically optimized for linearity. Thus, there will be no benefit in decreasing resistance (thereby increasing I_{CCQ}).

** With $M1 = 0$ Ω , I_{SHDN} increases linearly from $V_{SHDN}/I_{SHDN} = 1.8V/82\mu A$ to $4.2V/489\mu A$.

Said linear relationship can be used to scale M1 for higher V_{SHDN} voltage: use pin condition $V_{SHDN_pin}/I_{SHDN} = 2.4V/178\mu A$.

Calculate M1 for $V_{SHDN}/I_{SHDN} = 5V/178\mu A$: $M1 = (5-2.4)/(0.000178) = 14.6$ k Ω .



Caution! ESD Sensitive Device.

Exceeding Absolute Maximum Rating conditions may cause permanent damage.

Note: For additional information, please refer to [Manufacturing Note MN-001 — Package and Manufacturing Information](#).



All Guerrilla RF products are provided in RoHS compliant lead (Pb)-free packaging requiring no exemptions. Additional information for this topic can be found at this link - [Environmental and Restricted Substance Statement Library](#)

Recommended Operating Conditions

Parameter	Symbol	Specification			Unit	Condition
		Min.	Typ.	Max.		
Supply Voltage	V_{CC}	4	5	5.5	V	
Operating Temperature (package base)	$T_{PKG\ BASE}$	-40		105	°C	
RF Frequency Range	F_{RF}	2.2		2.5	GHz	Typical application schematic using the 2.3 to 2.5 GHz tuning set (note 2).
	F_{RF}	2.5		2.7	GHz	Typical application schematic using the 2.5 to 2.7 GHz tuning set (note 2).
RF_IN Port Impedance	Z_{RFIN}		50		Ω	Single-ended with 2-element match.
RF_OUT Port Impedance	Z_{RFOUT}		50		Ω	Single-ended with 2-element match.

Note 2: Contact the Guerrilla RF Applications team for guidance on optimizing the tuning of the device for alternative bands.

Nominal Operating Parameters – General

The following conditions apply unless noted otherwise: typical application schematic using the 2.5 to 2.7 GHz tuning set. $V_{CC} = +4.75$ to $+5.25$ V, $V_{SHDN} = \text{LOW}$, $I_{CCQ} = 150$ mA, $P_{OUT} = +23$ dBm, $F_{TEST} = 2.6$ GHz, $M2 = 6810 \Omega$, $M4 = 4750 \Omega$, $M10 = 1200 \Omega$, 50Ω system impedance, $T_{PKG \text{ BASE}} = -40$ to 105 °C. Typical values are at $V_{CC} = 5$ V, $I_{CCQ} = 150$ mA, $P_{OUT} = +23$ dBm, $F_{TEST} = 2.6$ GHz, $T_{PKG \text{ BASE}} = 25$ °C. Evaluation board losses are included within the specifications.

Parameter	Symbol	Specification			Unit	Condition
		Min.	Typ.	Max.		
Supply Quiescent Current	I_{CCQ}		150		mA	$I_{CCQ1} + I_{CCQ2}$. No RF applied.
Supply Current with RF Applied	I_{CC}		250		mA	$I_{CC1} + I_{CC2}$. RF applied with $P_{OUT} = +23$ dBm.
Enable Current 1	$I_{ENABLE1}$		0.2		mA	$V_{CC} = 5$ V.
Enable Current 2	$I_{ENABLE2}$		0.35		mA	$V_{CC} = 5$ V.
Enable Current 3	$I_{ENABLE3}$		0.8		mA	$V_{CC} = 5$ V.
Operating Temperature Range	$T_{PKG \text{ BASE}}$	-40		105	°C	Measured on package base.
Logic Input Low	V_{IL}	0		0.9	V	Applies to V_{SHDN} Input.
Logic Input High	V_{IH}	1.7		V_{CC}	V	Applies to V_{SHDN} Input.
Logic Current Low	I_{IL}		3		nA	Applies to V_{SHDN} Input, $V_{IL} = 0.9$ V.
Logic Current High	I_{IH}		80		μA	Applies to V_{SHDN} Input, $V_{IH} = 1.8$ V.
			320			Applies to V_{SHDN} Input, $V_{IH} = 3.3$ V.
Switching Rise Time	T_{RISE}		80		ns	Applies to V_{SHDN} Input.
Switching Fall Time	T_{FALL}		10		ns	Applies to V_{SHDN} Input.

Disabled Mode

Supply Quiescent Current	$I_{CCQ\text{-}SHDN}$		240		μA	$V_{CC} = 5$ V, $V_{SHDN}/V_{EN1}/V_{EN2}/V_{EN3} = \text{HIGH}$.
Enable Current 1	$I_{ENABLE1\text{-}SHDN}$		0.45		mA	$V_{CC} = 5$ V, $V_{SHDN}/V_{EN1}/V_{EN2}/V_{EN3} = \text{HIGH}$.
Enable Current 2	$I_{ENABLE2\text{-}SHDN}$		0.8		mA	$V_{CC} = 5$ V, $V_{SHDN}/V_{EN1}/V_{EN2}/V_{EN3} = \text{HIGH}$.
Enable Current 3	$I_{ENABLE3\text{-}SHDN}$		3.4		mA	$V_{CC} = 5$ V, $V_{SHDN}/V_{EN1}/V_{EN2}/V_{EN3} = \text{HIGH}$.

Thermal Data (Stage 1 and Stage 2)

See plot of junction temperature vs. output power.						On standard evaluation board.
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Nominal Operating Parameters – RF: 2.5 to 2.7 GHz, 5 V

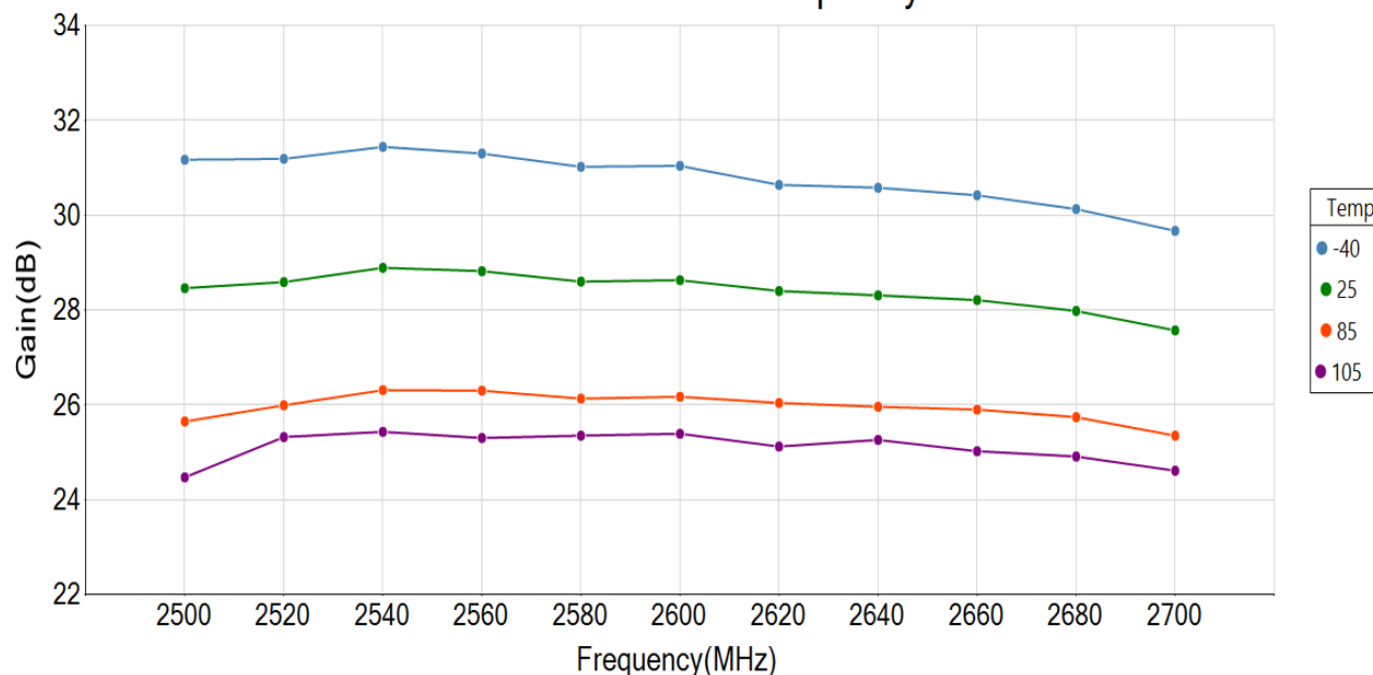
The following conditions apply unless noted otherwise: typical application schematic using the 2.5 to 2.7 GHz tuning set. $V_{CC} = +4.75$ to $+5.25$ V, $V_{SHDN} = \text{LOW}$, $I_{CCQ} = 150$ mA, $P_{OUT} = +23$ dBm, $F_{TEST} = 2.6$ GHz, $M2 = 6810 \Omega$, $M4 = 4750 \Omega$, $M10 = 1200 \Omega$, 50Ω system impedance, $T_{PKG \text{ BASE}} = -40$ to 105 °C. Typical values are at $V_{CC} = 5$ V, $I_{CCQ} = 150$ mA, $P_{OUT} = +23$ dBm, $F_{TEST} = 2.6$ GHz, $T_{PKG \text{ BASE}} = 25$ °C. Evaluation board losses are included within the specifications.

Parameter	Symbol	Specification			Unit	Condition
		Min.	Typ.	Max.		
Small Signal Gain	S21	27	28.5		dB	LTE 20MHz 100RB TM1.1 Downlink Waveform with 9.6 dB PAR, $F_{TEST} = 2.6$ GHz, $V_{CC} = 5$ V, $P_{IN} = -25$ dBm.
Standby Mode Gain	S21 _{STBY}		-33.5		dB	Disabled Mode, LTE 20MHz 100RB TM1.1 Downlink Waveform with 9.6 dB PAR, $V_{SHDN}/V_{EN1}/V_{EN2} = \text{HIGH}$, $P_{IN} = 0$ dBm.
Input Return Loss	S11		< -7		dB	$F_{RF} = 2.5$ to 2.7 GHz
Output Return Loss	S22		< -5		dB	$F_{RF} = 2.5$ to 2.7 GHz
Reverse Isolation	S12		< -40		dB	$F_{RF} = 2.5$ to 2.7 GHz
Noise Figure	NF		3		dB	On standard evaluation board.
Output 3rd Order Intercept Point	OIP3		45		dBm	+23 dBm P_{OUT} per tone using 600 kHz spacing.
Output 1 dB Compression Power	OP1dB		32		dBm	$V_{CC} = 5$ V. Sine wave input.
Adjacent Channel Leakage Ratio	ACLR		-45		dBc	$P_{OUT} = +23$ dBm, LTE 20MHz 100RB TM1.1 Downlink Waveform with 9.6 dB PAR, $F_{TEST} = 2.6$ GHz, $V_{CC} = 5$ V.

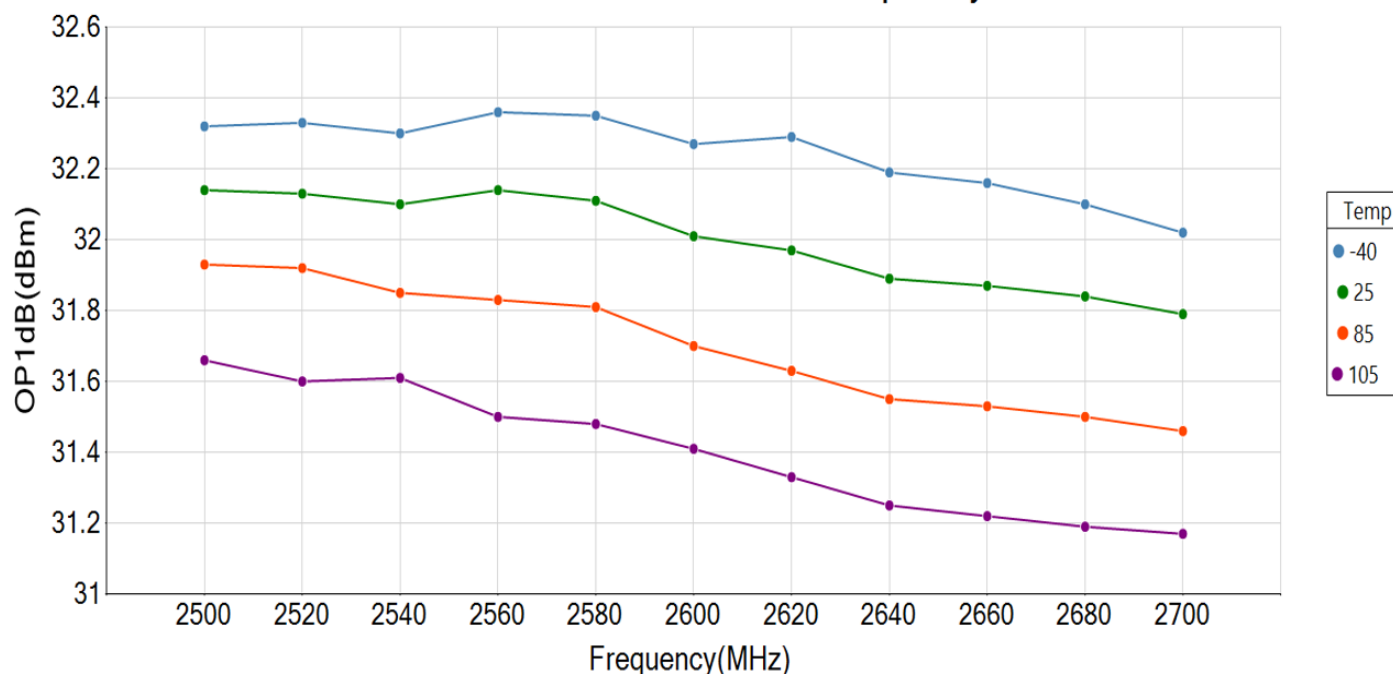
Note 3: Min/Max limits defined using *modelled estimates* that account for part-to-part variations and expected process spreads. As additional production lots are fabricated, accumulated test data will be used to refine the Min/Max limits.

GRF5526 Typical Operating Curves: 2.5 to 2.7 GHz Tune

GRF5526 Gain vs Frequency

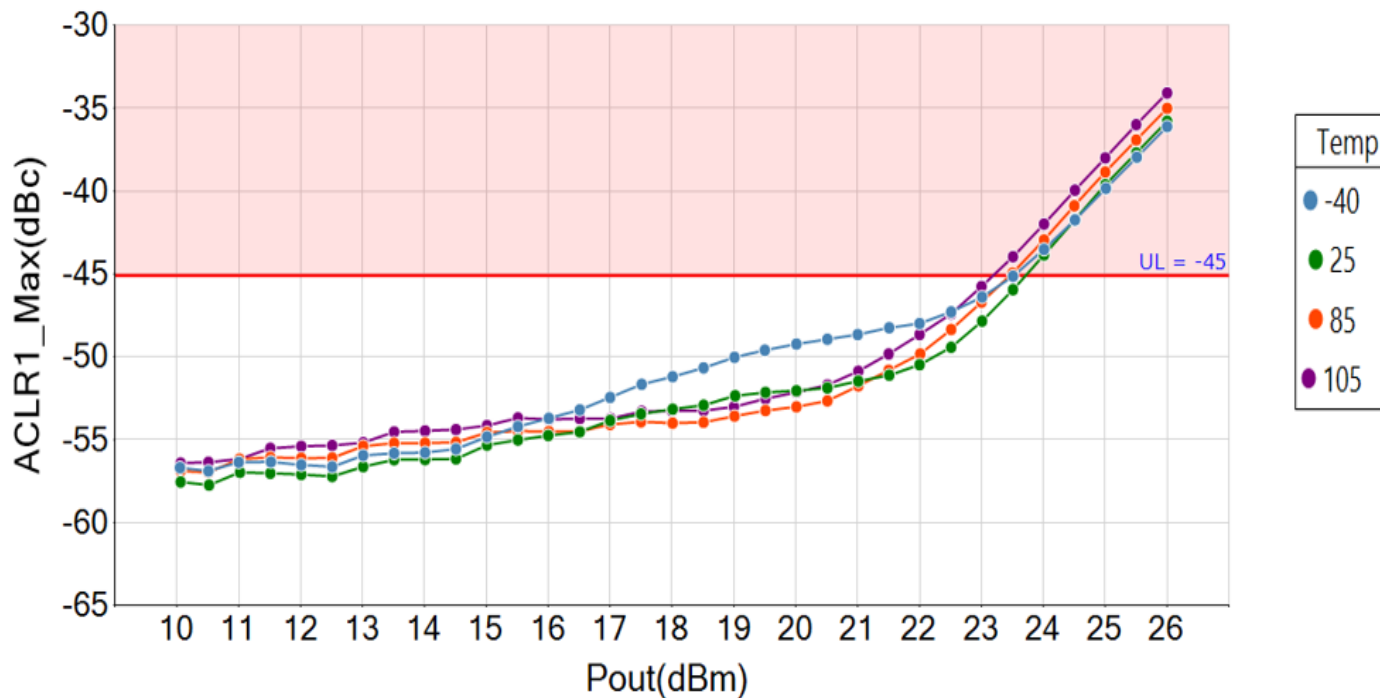


GRF5526 OP1dB vs Frequency



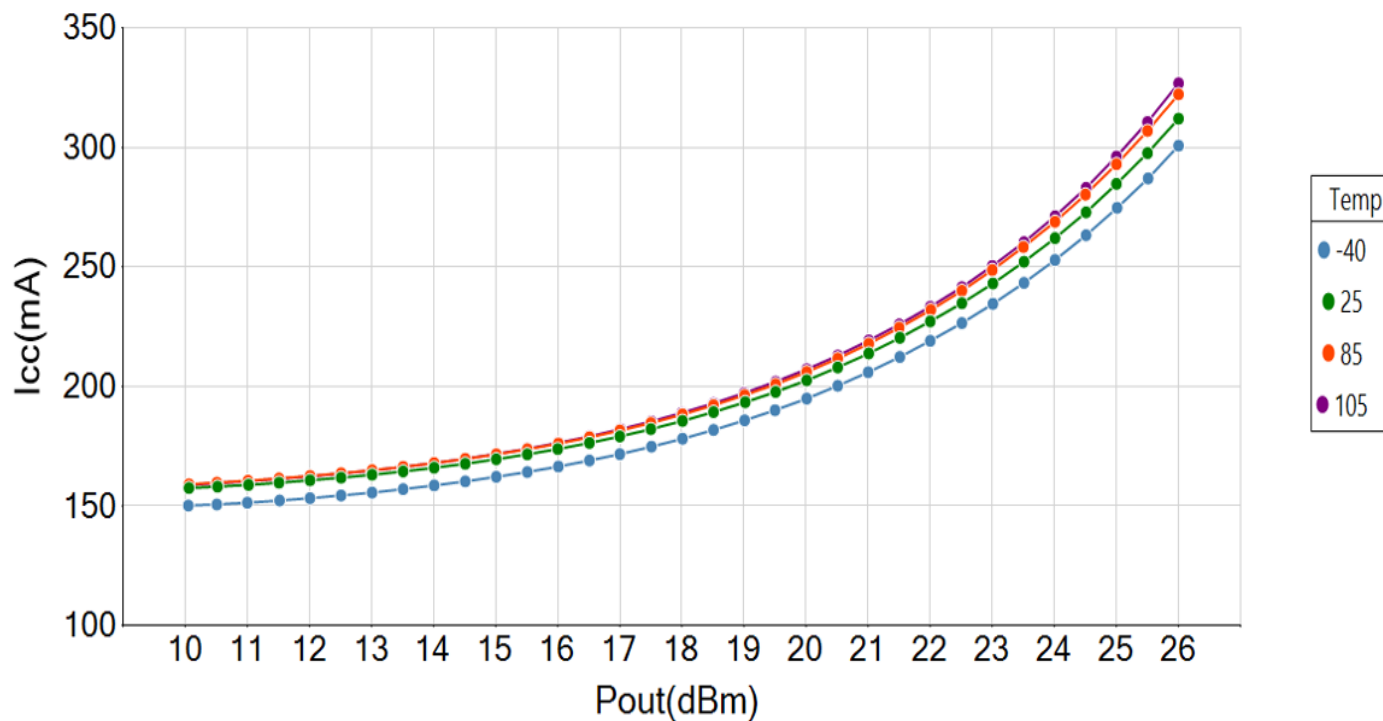
GRF5526 Typical Operating Curves: ACLR vs. P_{OUT} (LTE 20Q100RB TM1.1 9.6 dB PAR)

GRF5526 ACLR1 vs Pout at Frequency = 2570 MHz and Modulation = LTE_20MHz_100RB



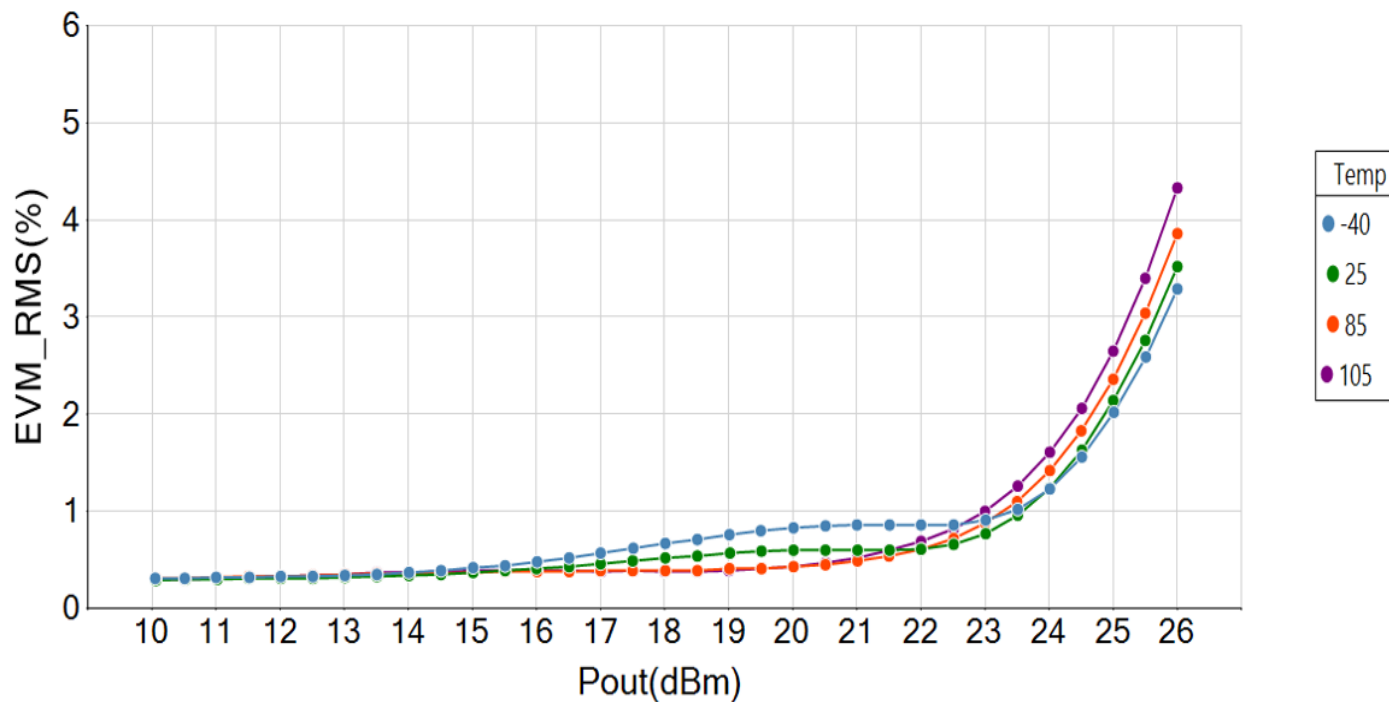
GRF5526 Typical Operating Curves: Stage 1 + Stage 2 I_{CC} vs. P_{OUT} (LTE 20Q100RB TM1.1 9.6 dB PAR)

GRF5526 I_{CC} vs P_{OUT} at Frequency = 2570 MHz and Modulation = LTE_20MHz_100RB

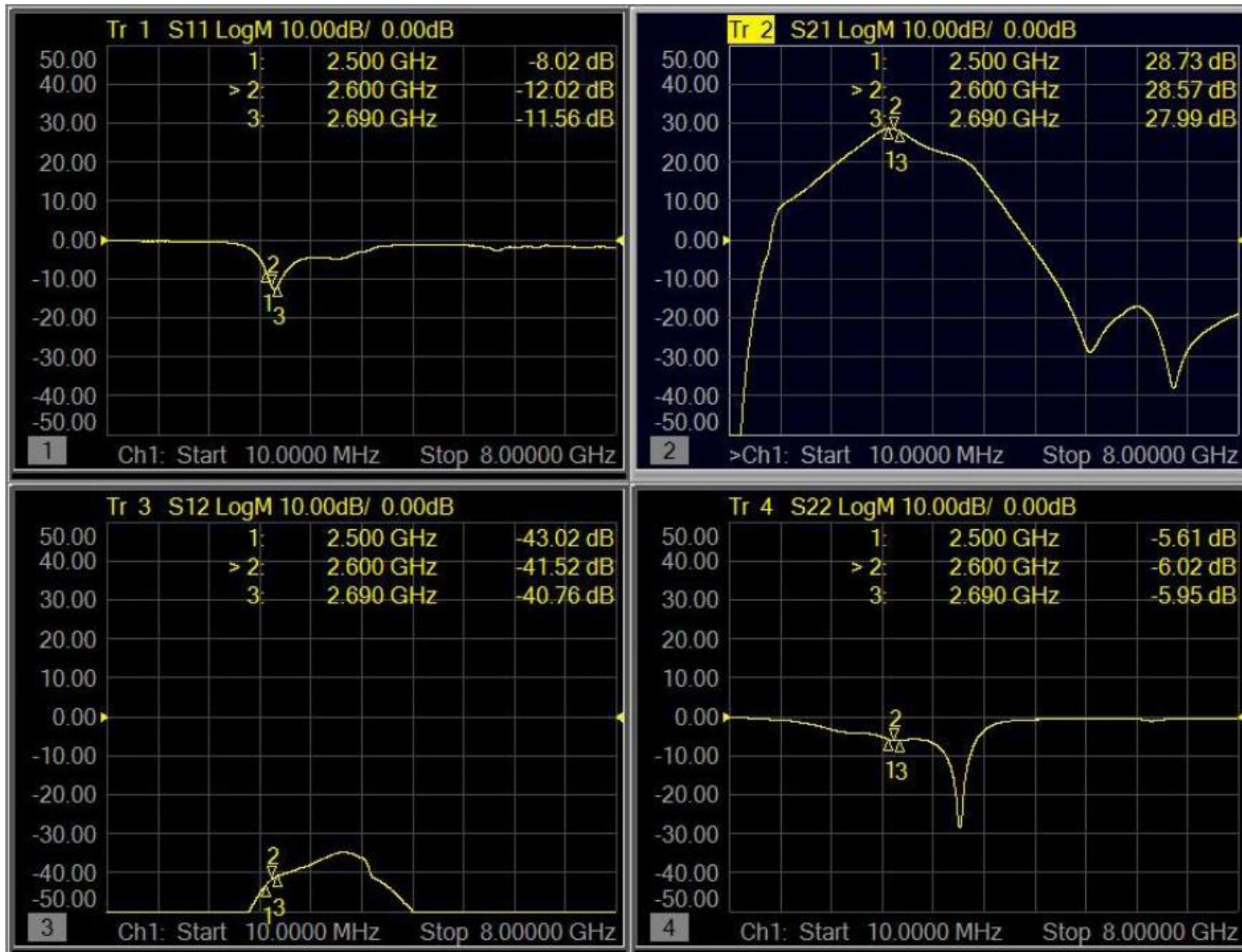


GRF5526 Typical Operating Curves: EVM vs. P_{OUT} (LTE 20Q100RB TM1.1 9.6 dB PAR)

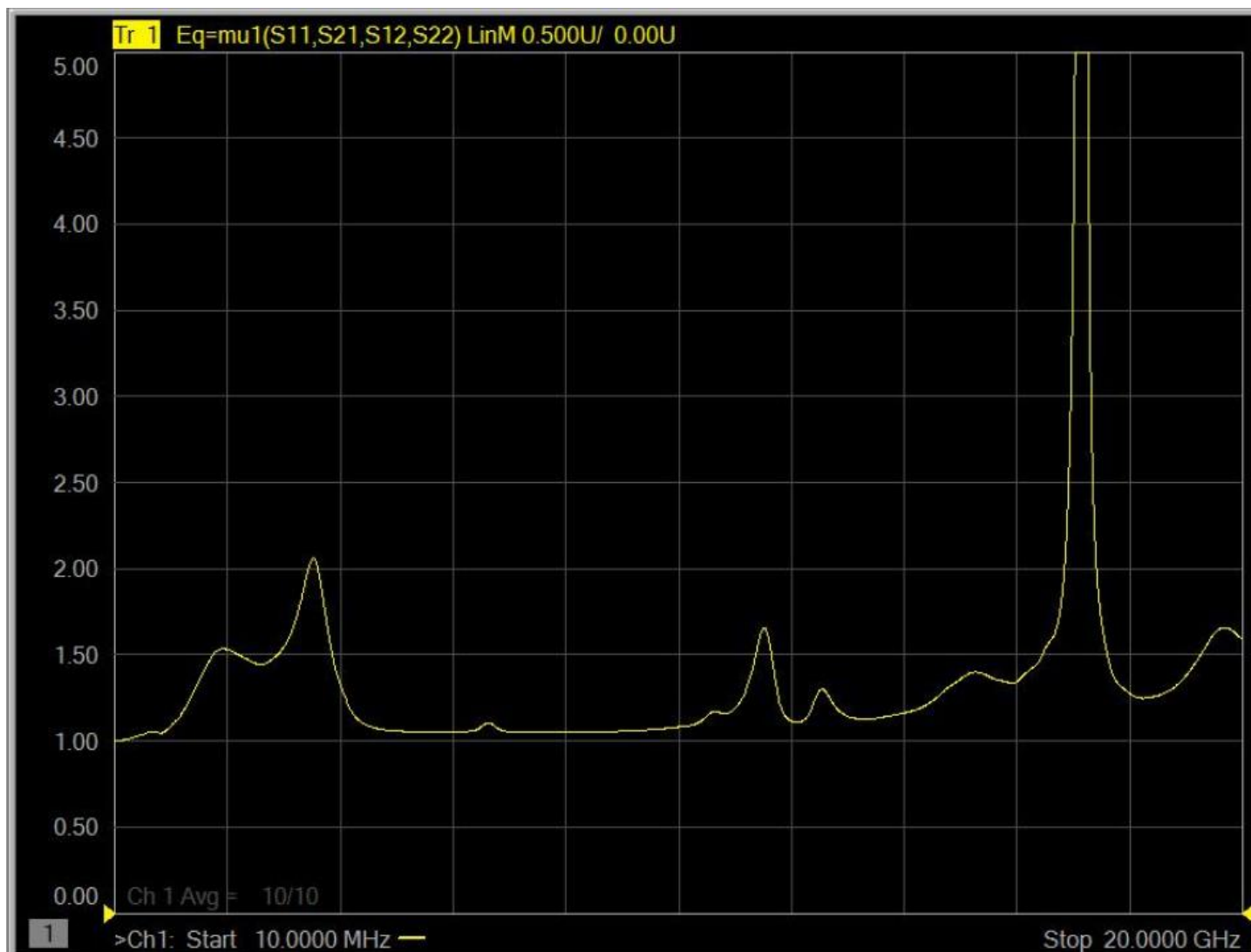
GRF5526 EVM_RMS vs Pout at Frequency = 2570 MHz and Modulation = LTE_20MHz_100RB



GRF5526 Typical Operating Curves: S-Parameters (2.5 to 2.7 GHz Tune)



GRF5526 Typical Operating Curves: Stability Mu (10 MHz to 20 GHz)



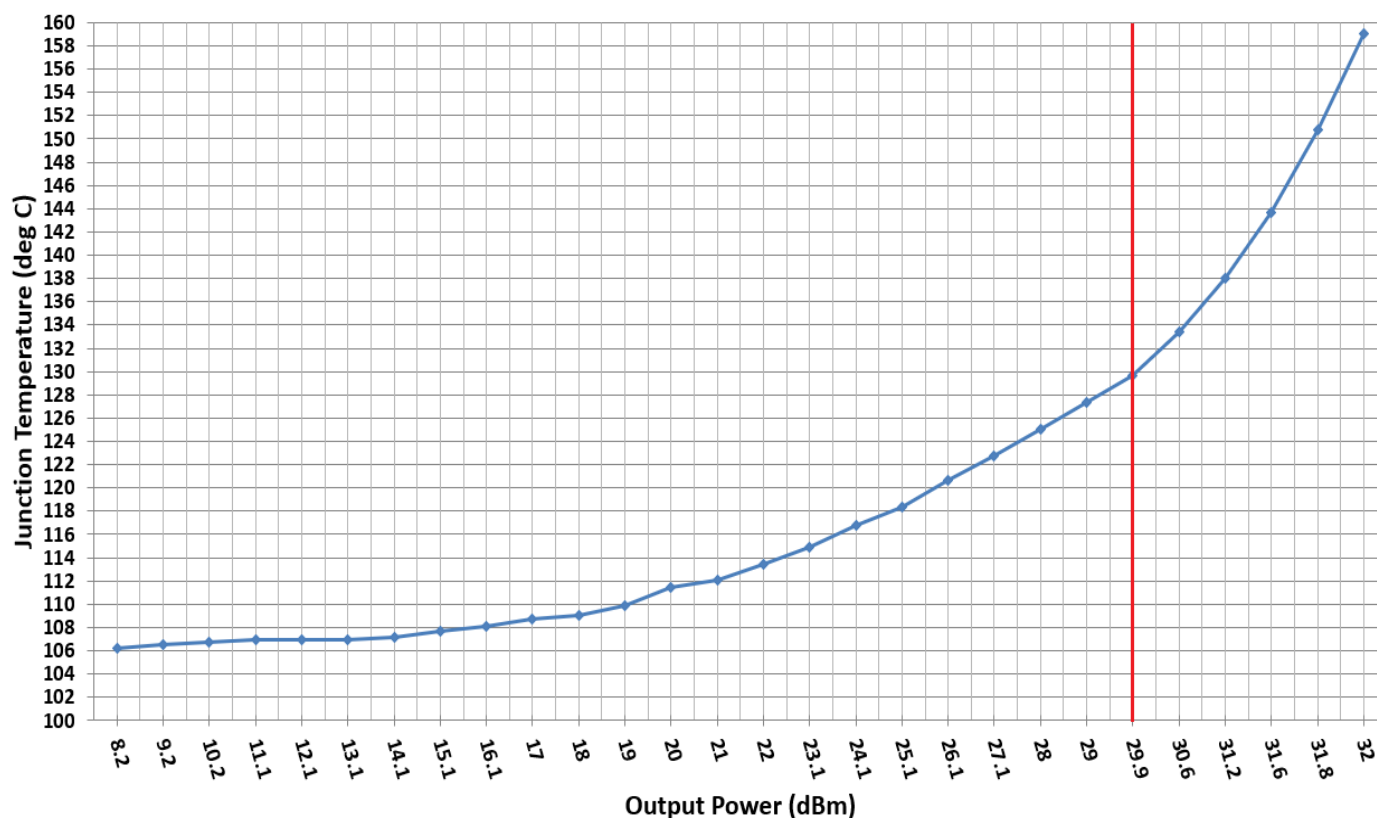
Note: Mu factor ≥ 1.0 implies unconditional stability

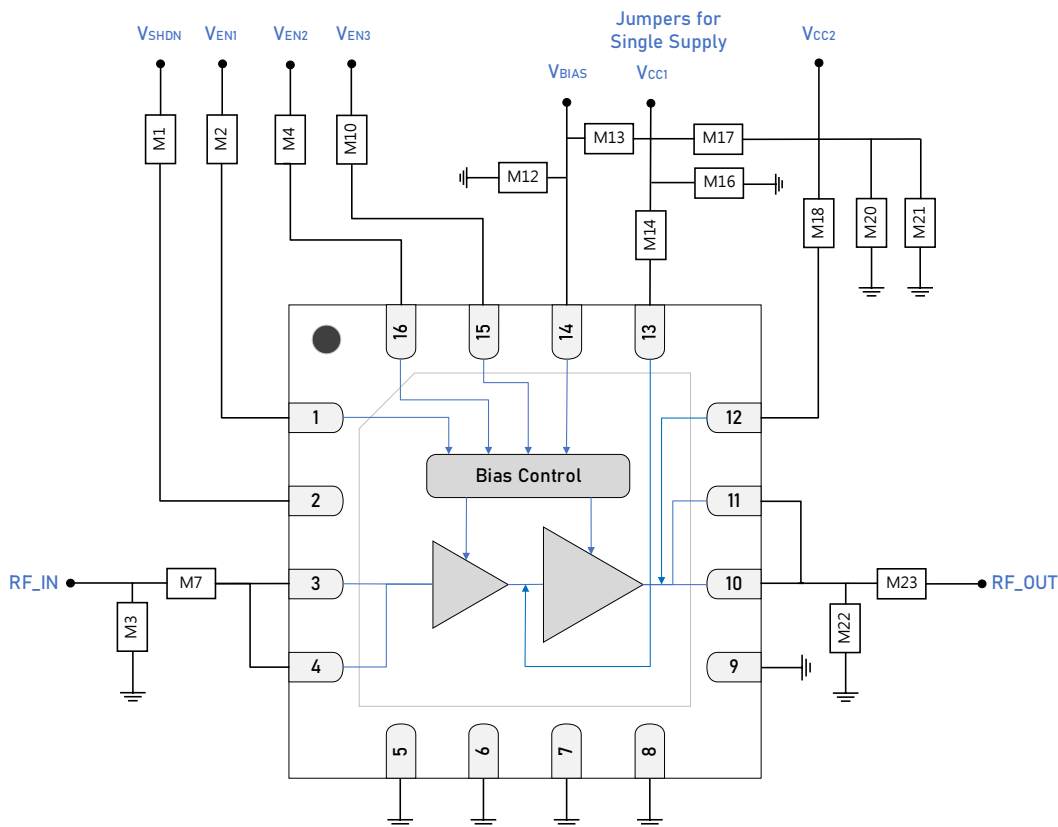
GRF5526 Typical Operating Curves: Junction Temperature (per application schematic @ 85 °C)

GRF5526, being a 2-stage device, sees one of the stages governing junction temperature over power sweep. Red line = 30 dBm shows where T_J is equivalent in both stages. At left of red line, stage 2 governs T_J ($Q_2 T_J$ is higher). To the right of red line, stage 1 governs T_J ($Q_1 T_J$ is higher).

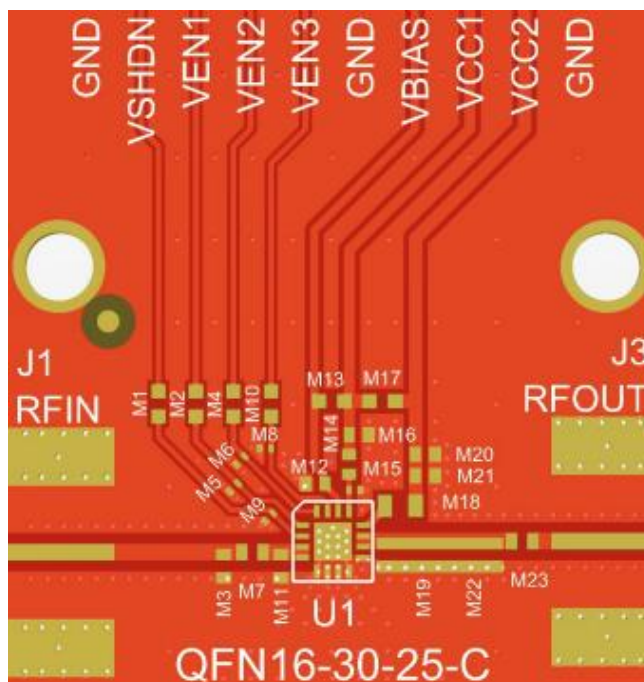
Setting bias resistor M2/M4 per application schematic ensures best linearity and yields thermal performance shown in the plot. If the application does not require high IMD3/ACLR linearity, bias resistor can be adjusted higher. This will lower bias point(s) and junction temperature will be contained within/below that shown in the plot.

GRF5526





GRF5526 Standard Evaluation Board Schematic

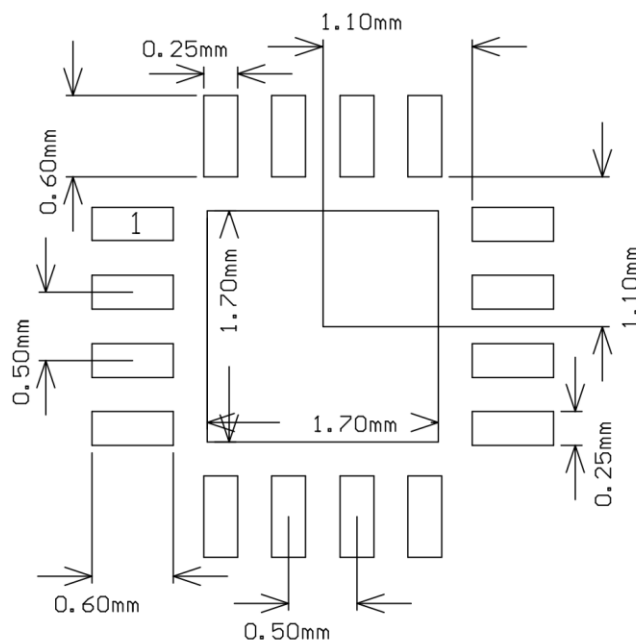


GRF5526 Evaluation Board Assembly Diagram

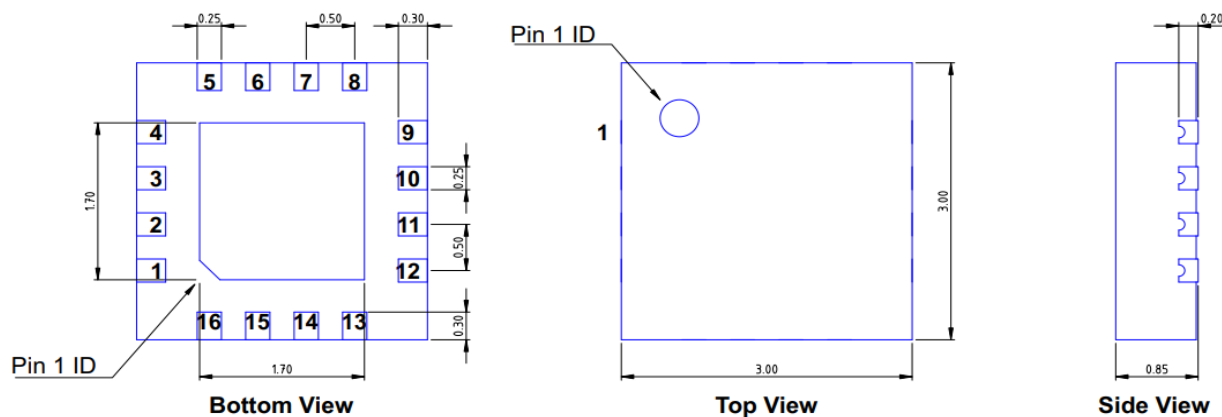
GRF5526 Evaluation Board Assembly Diagram Reference

Component	Type	Manufacturer	Family	Value	Package Size	Substitution
M1	Resistor	Various	--	0 Ω	0402	ok
M2	Resistor	Various	1%	6810 Ω	0402	ok
M3	Inductor	Murata	LQGWH	1.1 nH	0402	ok
M4	Resistor	Various	1%	4750 Ω	0402	ok
M7	Capacitor	Murata	GJM	1.3 pF	0402	ok
M10	Resistor	Various	1%	1200 Ω	0402	ok
M12	Capacitor	Murata	GRM	4.7 μ F	0402	ok
M13	Resistor (jumper)	Various	--	0 Ω	0402	ok
M14	Inductor	Murata	LQW	1.7 nH	0402	ok
M16	Capacitor	Murata	GRM	4.7 μ F	0402	ok
M17	Resistor (jumper)	Various	--	0 Ω	0402	ok
M18	Inductor	Murata	LQW	4.3 nH	0603	ok
M20	Capacitor	Murata	GRM	4.7 μ F	0402	ok
M21	Capacitor	Murata	GRM	100 pF	0402	ok
M22	Capacitor	Murata	GJM	1.3 pF	0402	ok
M23	Capacitor	Murata	GJM	10 pF	0402	ok
Evaluation Board	QFN16-30-25-C					

Note 4: Standard evaluation board bias: $V_{CC} = 5$ V, $V_{ENABLE} = 5$ V.



3 x 3 mm QFN-16 Suggested PCB Footprint (Top View)



QFN16 3x3mm
Dimensions in millimeters
Dimensional Tolerance: ± 0.05

3 x 3 mm QFN-16 Package Dimensions

Package Marking Diagram



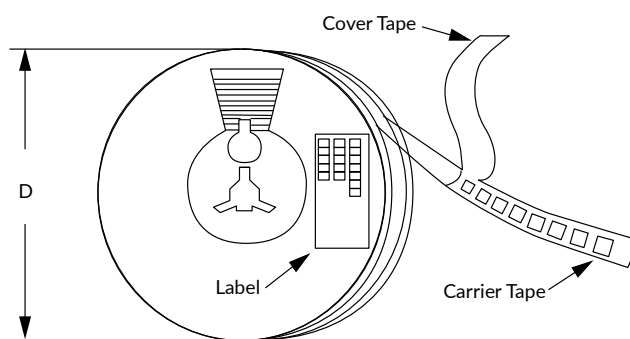
- Line 1: "YY" = YEAR. "WW" = WORK WEEK the device was assembled.
- Line 2: "GRF" = Guerrilla RF.
- Line 3: "XXXX" = Device PART NUMBER.

Tape and Reel Information

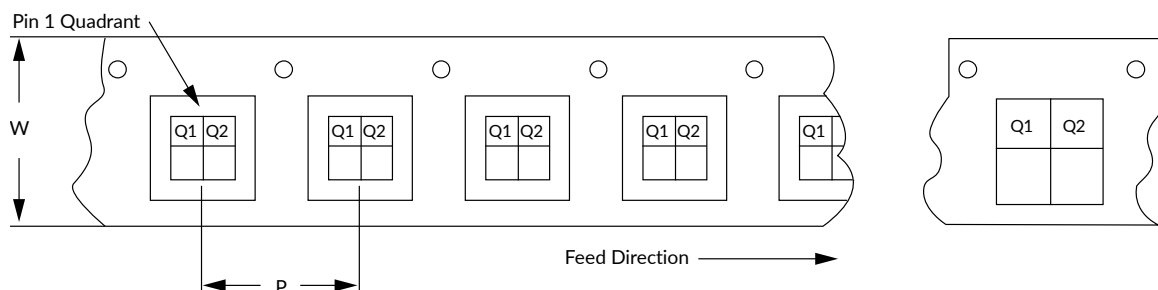
Guerrilla RF's tape and reel specification complies with Electronic Industries Alliance (EIA) standards for "Embossed Carrier Tape of Surface Mount Components for Automatic Handling" (reference EIA-481). See the following page for the Tape and Reel Specification and Device Package Information table, which includes units per reel.

Devices are loaded with pins down into the carrier pocket with protective cover tape and reeled onto a plastic reel. Each reel is packaged in a cardboard box. There are product labels on the reel, the protective ESD bag and the outside surface of the box.

For the Tape and Reel Reference Table, please refer to: <https://www.guerrilla-rf.com/prodFiles/Manufacturing/MN001.pdf>



Tape and Reel Packaging with Reel Diameter Noted (D)



Carrier Tape Width (W), Pitch (P), Feed Direction and Pin 1 Quadrant Information

Revision History

Revision Date	Description of Change
December 16, 2021	Preliminary Data Sheet.
February 14, 2022	Updated Pin Assignments table.
May 4, 2022	Changed RF Input Power (Pin Max) from TBD to 23 dBm.
August 25, 2022	Upgraded Evaluation Board from RevB to RevC. On Evaluation Board BOM and Schematic: changed C1 to M12. On Evaluation Board BOM: changed M3 from 1.6 to 1.1 nH (LQG to LQGWH) and M7 from 1.4 to 1.3 pF.
October 3, 2022	Added Thermal Resistance specification.
November 1, 2022	Added Characterization Plots. Added new S-parameter and Mu Plots.
January 27, 2023	In Absolute Ratings Table: Added the following condition to Maximum Dissipated Power for Stage 1 & 2: DC only (no RF applied). Changed Stage 1 Maximum Dissipated Power from TBD to 500 mW. Changed Stage 2 Maximum Dissipated Power from TBD to 1400 mW.
June 6, 2023	Release Ø Data Sheet.
September 19, 2023	Release A Data Sheet.
January 17, 2025	Updated Data Sheet with cosmetic changes only. No change to device or device specifications.
May 12, 2025	Extended lower frequency range from 2.3 to 2.2 GHz. Updated frequency of family of part numbers listed on page 1.



Data Sheet Classifications

Data Sheet Status	Notes
Advance	S-parameter and NF data based on EM simulations for the fully packaged device using foundry-supplied transistor S-parameters. Linearity estimates based on device size, bias condition and experience with related devices.
Preliminary	All data based on limited evaluation board measurements taken within the Guerrilla RF Applications Lab. All parametric values are subject to change pending the collection of additional data.
Release Ø	All data based on measurements taken with <i>production-released</i> material. TYP values are based on a combination of ATE and bench-level measurements, with MIN/MAX limits defined using <i>modelled estimates</i> that account for part-to-part variations and expected process spreads. Although unlikely, future refinements to the TYP/MIN/MAX values may be in order as multiple lots are processed through the factory.
Release A-Z	All data based on measurements taken with production-released material <i>derived from multiple lots which have been fabricated over an extended period of time</i> . MIN/MAX limits may be refined over previous releases as more statistically significant data is collected to account for process spreads.

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